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METHOD AND APPARATUS FOR COMMUNICATING SIGNALS

FIELD OF THE INVENTION

This invention relates to communication of signals in a time division multiplexed (TDM) system and, more particularly, to a method and apparatus that improves operation of a TDM system, such as by preventing an output at a receiver thereof from containing undesired noise patterns for periods long enough to cause a misoperation of a receiver output. The invention has application, for example, to teleprotection systems.

BACKGROUND OF THE INVENTION

Audio Tone Teleprotection is a method of communicating binary states from one substation to another over audio telecommunications links. These binary states are used, for example, to control power line circuit breakers and generators. This practice

has existed for over 30 years and has principally been applied over entirely analog systems such as Frequency Division Multiplexers, Analog Microwave, and Direct Analog Fiber.

When a signal occurs in any of the above media, noise occurs at the input of the teleprotection and circuitry inside the teleprotection is designed to prevent a "misoperation". A misoperation is a false state on one of the binary outputs. The result of a false output can be disastrous, e.g. blackouts, fires, and/or equipment damage. It is important to prevent these misoperations in the event of loss of signal, since systems like digital microwave are frequently susceptible to loss of signal, usually due to adverse weather conditions.

With the recent proliferation of telecommunications has come the practice of applying audio teleprotection over Time Division Multiplexed (TDM) digital communications networks. Under failure conditions, different noise patterns occur at the input of the teleprotection. These noise patterns can very closely mimic valid signals indicating incorrect states, possibly causing misoperation of the teleprotection equipment.

The most common form of TDM used in teleprotection is known as T1 or DS-1. T1 is a format defined by the Bell system for digital transmission of analog signals. It is designed to carry 24 audio channels over a 1.544 Mb/s serial digital link, as illustrated in Figure 1. As shown in Figure 2, one frame has 24 8-bit words (192 bits) plus 1 framing bit, so one out of every 193 bits is used to create a framing pattern. T1 frames are transmitted at the rate of 8000 per second, so, with 193 bits per frame, the bit rate is 1.544 megabits per second. This framing pattern repeats every 1.5 or 3.0 ms,

depending on whether Superframe or Extended Superframe is employed. Superframe is a 1.5 ms frame consisting of 12 consecutive T1 frames. Since standard framing practice is to look for at least 2 bad framing patterns before declaring a loss of frame, the process can take at least 3 ms. Extended Superframe is made up of 24 frames and can take 6 milliseconds or more to detect a loss of frame.

With standard telephone grade multiplexers, a noisy signal is output from the voice channels for up to 2 seconds after a complete loss of signal.

It is among the objects of the present invention to provide a method and apparatus which overcomes disadvantages of existing approaches such as those described, and to prevent misoperations of teleprotection systems in the event of TDM signal loss. It is among the further objects of the invention to achieve this in a way that is compatible with existing teleprotection design, due to the large installed base of teleprotection equipment.

Further features and advantages of the invention will become more readily apparent from the following detailed description when taken in conjunction with the accompanying drawings.

BRIEF DESCRIPTION OF THE DRAWINGS

Figure 1 is a diagram illustrating the T1 (or DS-1) time division multiplexed (TDM) format commonly used for digital transmission of signals.

Figure 2 is a further diagram illustrating a frame of T1 format.

Figure 3 is a block diagram of a transmitter in accordance with an embodiment of the invention, and which can be used in practicing the method of the invention.

Figure 4 is a block diagram of a receiver in accordance with an embodiment of the invention, and which can be used in practicing the method of the invention.

Figure 5 is a functional block diagram of an embodiment of the transmitter.

Figure 6 is a further diagram of the receiver operation.

Figure 7 is a flow diagram that represents an embodiment of the programmable logic for the decision engine of the Figure 6 embodiment.

DETAILED DESCRIPTION

Referring to Figure 3, there is shown a block diagram of the transmitter portion of an embodiment of the invention. The block 310 represents a T1 framer circuit, adapted, in a manner to be described, to produce a special framing pattern (dashed block 312). In the example of Figure 3, a number of inputs, which can for example be teleprotection signals at audio frequencies, are input, via interfaces 331, 332, 333......, and the transmit bit stream lines 340, to the T1 framer circuit 310. Each of the interfaces 331, 332, 333 receives frame timing signals from the framer circuit 312, and this provides timing for the bit stream input to the framer circuit. The framer circuit 312 inserts the framing signals, and produces the time division multiplexed (TDM) output signal (T1 output signal on line 360).

input to selector switch 423 is frame timing signals output from framing pattern detector 421 on line 421a. The control input to selector switches 415 and 423 is the output of decision engine 650 on line 421b. The decision engine 650 receives the in-frame signal output of the T1 deframer circuit, and a second signal from framing pattern detector, this signal being present when the framing pattern detector detects that the received bit stream is "in frame". The outputs of the selector switches 415 and 423, respectively, on lines 415a and line 423a, are also coupled to respective data inputs and frame timing inputs of the audio interfaces 431, 432, 433,

Referring to Figure 5, there is shown a functional block diagram of the transmitter first shown in Figure 3. In this representation, the transmitter includes a series of low speed modules, which output data into a common backplane in response to timing signals generated by the timing generator 512 that is part of the T1 framer circuit (block 310 of Figure 3) which typically can be implemented as an integrated circuit. Those modules put out data in their assigned time slots (1-23) and the framer circuit receives the backplane data and inserts it into the T1 frame via gate 515. The special framing pattern is stored in the register 513 inside the framer chip. In the present embodiment, the framer is programmed to insert that pattern in the last timeslot of the T1 frame (24). The bit stream and the associated framing pattern are received by the frame generator 520, and the output T1 transmitted signal is output, on line 360, via the T1 line interface unit 530.

Referring to Figure 6, there is shown a representation of the receiver circuit 420 (of Figure 4), which, in this embodiment, is implemented using programmable logic.

The incoming T1 serial data is shifted into a register 605 that is one frame plus 8 bits

(201 bits) long. The selected fast reframe pattern (01010011 in this example) is stored in memory 610. Comparators 620 and 630 operate to compare the first 8 bits and the last 8 bits of register 605 to the stored pattern. When the first and last 8 bits of this register both match the stored framing pattern, a match is declared. This condition is input (on lines 620a and 630a) to a decision engine 650, which also receives the in frame signal from T1 deframer and which can declare an out of frame condition quickly after seeing the match signal go away. The decision engine uses the standard in frame signal from the T1 deframer and the in frame signal from the special framing detector and decides what framing signals and what data to supply to the audio interfaces. It does this via control signals 421B.

Typically, the framer circuit generates frame-timing signals to the individual channels alerting them to the proper start of frame. When the system is out of frame these signals occur at the wrong time. In the present embodiment, however, the circuit of Figures 4 and 6 looks for the special framing signal and, if it sees two matching 01010011 patterns exactly 125 microseconds apart, produces an output that declares an "in frame" condition. It now intercepts the frame-timing signals from the framer and inserts its own, proper, frame-timing signals. These signals allow the channels to properly operate despite the framer circuit still being out of frame.

When a loss of incoming signal occurs, the special framing circuit does not see the predetermined frame pattern repeat twice and produces an output that declares a loss of frame. This will happen between 30 and 250 microseconds after a loss of signal. At that time, it intercepts the data from the framer to the channels and inserts a stream of logical ones. This eliminates any audio signal from being generated. Since

the audio interface converts the bit stream to an analog signal by converting each 8 bits of time slot data into an analog quantity, a steady stream of ones would convert to a steady state DC level. Further, since DC is not in the audio band, no output would be generated. Since the noise can occur only from the time of loss of signal to the time the logical ones are inserted, it lasts much less than 1 millisecond. Audio teleprotection channels currently in use will not misinterpret 1 millisecond of noise to produce a false output state (that is, misoperate), so a vexing problem in the art is solved.

Referring to Figure 7, there is shown a flow diagram that represents an embodiment of the programmable logic routine for the decision engine 650. The blocks 710 and 720 implement an initial capture phase of the routine (whereby, after initialization or after loss of sync, an in-frame state can be established), and the blocks 730 and 750 implement a normal operation phase of the routine (after an in-sync state is established, and operation is monitored for a possible loss of sync). The decision block 710 represents determination of whether two fast reframe patterns, one data frame apart, have been received. If not, the block 710 is re-entered, for continuation of this monitoring. If so, an in-frame state is declared (block 720) which results in production of an in-frame signal on line 421b. During the next phase of the routine, the block 730 represents monitoring for loss of the fast reframe pattern, and this continues (re-entry from the "no" output) until such time as a loss of pattern is detected. In the present embodiment, one loss of the fast reframe pattern is permitted. However, if two successive fast reframe patterns are not present (i.e., in Figure 6, both comparator outputs indicating no match), the loss of sync state is declared (block 750), and the signal on line 421b is changed so that, in Figure 4, the selector switches 415 and 423

will change the signals they select for coupling to audio interfaces, as previously described. As seen in Figure 6, the decision engine 650 can also receive, for processing, the in-frame signal from the T1 deframer

The invention has been described with reference to particular preferred embodiments, but variations within the spirit and scope of the invention will occur to those skilled in the art. For example, it will be understood that other means and techniques can be utilized for generating and sensing the fast reframe pattern, and for implementing the auxiliary timing signals and the override information signals.